

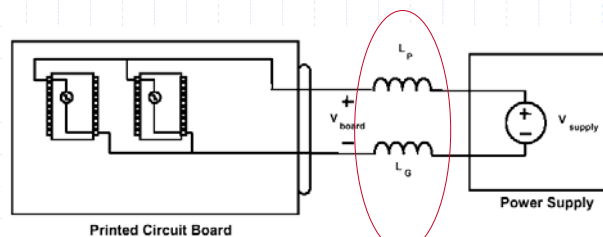


EMC Considerations for DC Power Design

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National Sun Yat-sen University



Power Bus Noise below 5MHz

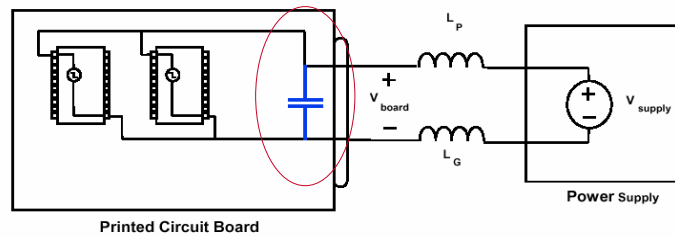


Power Distribution Model ~ (0 - 5 MHz)

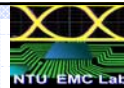


Power Bus Noise below 5MHz (Solution)

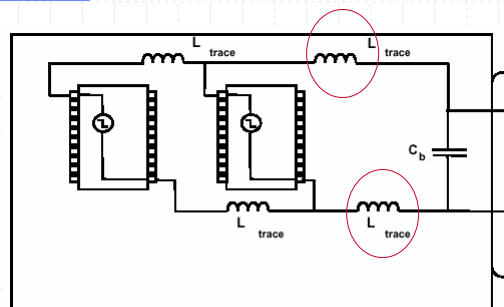
Add Bulk capacitance



Power Distribution Model ~ (0 - 5 MHz)

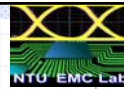


Power Bus Noise between 5MHz to 50MHz

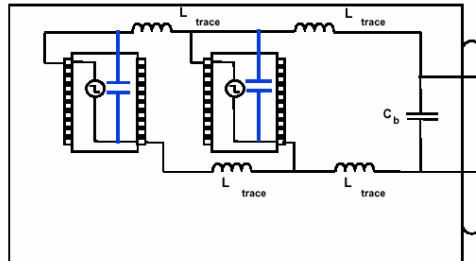


Power Distribution Model ~ (5 - 50 MHz)

Board without power and ground planes

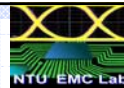


Power Bus Noise between 5MHz to 50MHz (Solution)

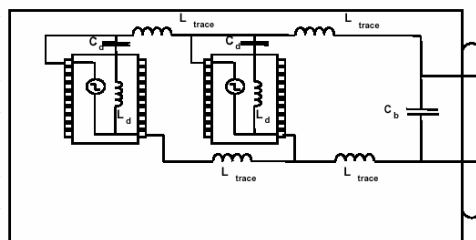


Power Distribution Model ~ (5 - 50 MHz)

Board without power and ground planes

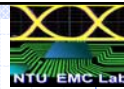


Power Bus Noise between 50MHz to 500MHz

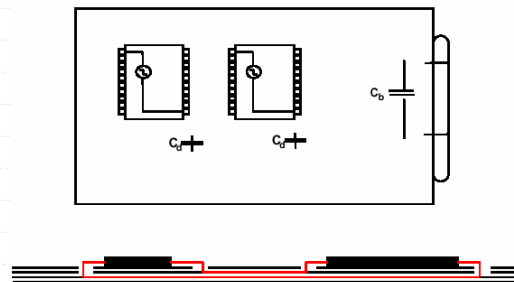


Power Distribution Model ~ (50 - 500 MHz)

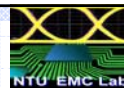
Board without power and ground planes



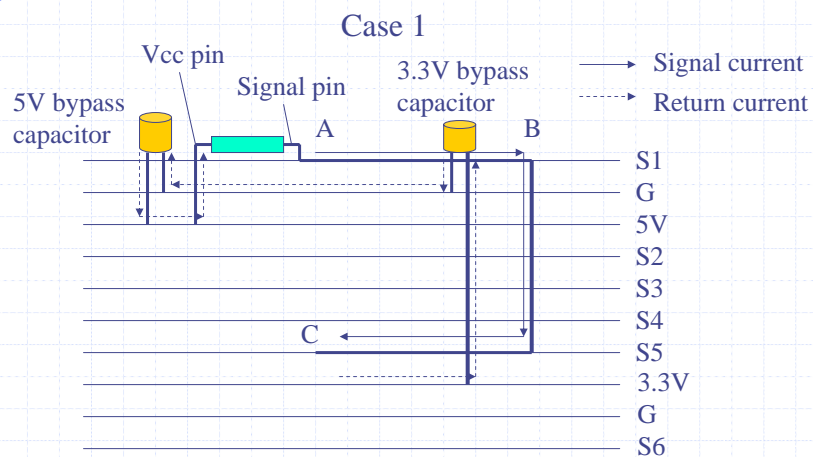
Power Bus Noise between 50MHz to 500MHz (Solution)



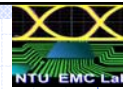
Power Distribution Model ~ (5 - 500 MHz)
Board with power and ground planes



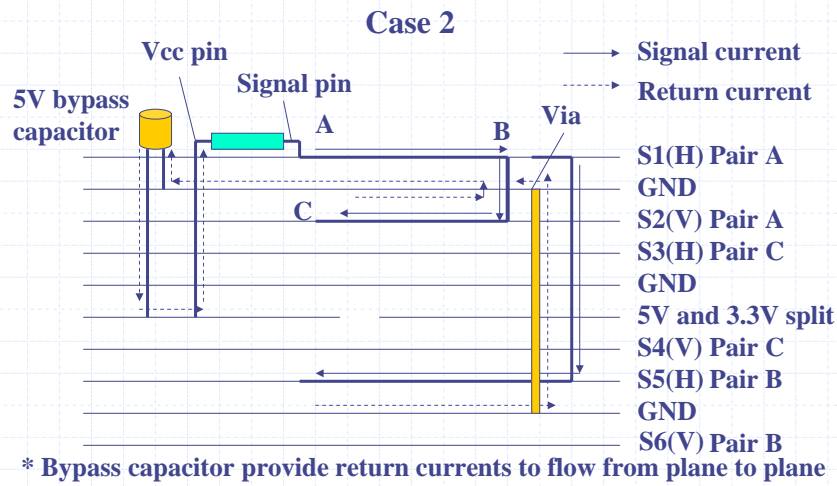
RF Path in power bus (I)



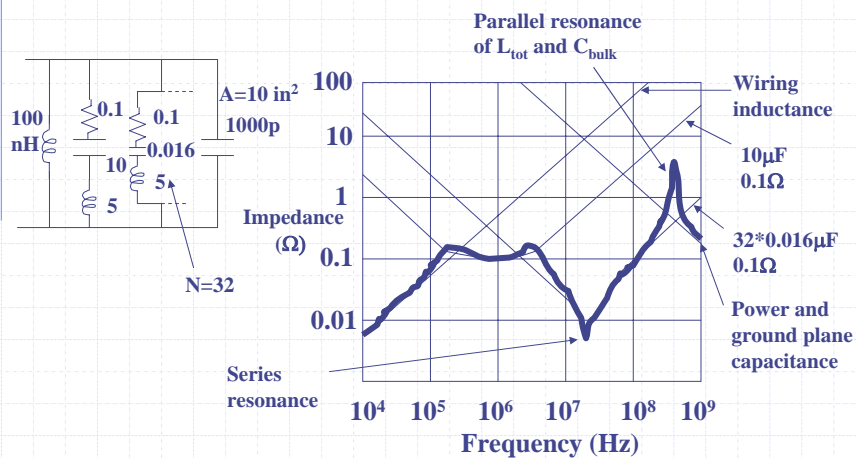
* Bypass capacitor provide return currents to flow from plane to plane

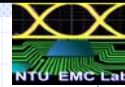


RF Path in power bus (II)



A simple power bus model





Power Bus Impedance

TABLE I
TARGET IMPEDANCE IS DROPPING BY A
FACTOR OF 5 EVERY COMPUTER GENERATION

year	voltage (volts)	power dissipated (watts)	current (amps)	Z _{target} (mOhms)	frequency (MHz)
1990	5.0	5	1	250	16
1993	3.3	10	3	54	66
1996	2.5	30	12	10	200
1999	1.8	90	50	1.8	600
2002	1.2	180	150	0.4	1200

$$Z_{\text{target}} = \frac{(\text{Power Supply Voltage}) \times (\text{allowed ripple})}{\text{current}}$$
$$= \frac{(5 \text{ V}) \times (5\%)}{1 \text{ A}} = 0.250 \text{ Ohms.}$$



Power Bus Impedance (an example)

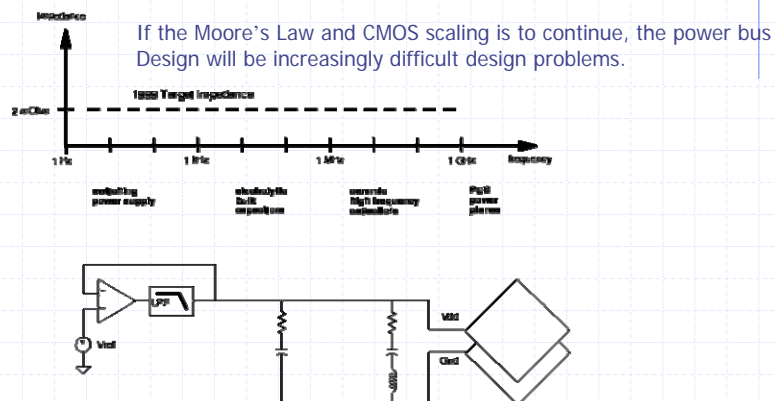
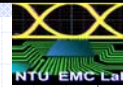
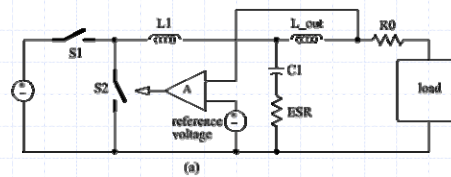


Figure 1: A flat power supply impedance vs. frequency is met by using VRM, bulk capacitor, high frequency ceramic capacitor and power plane components.



Voltage Regulator Module (VRM)

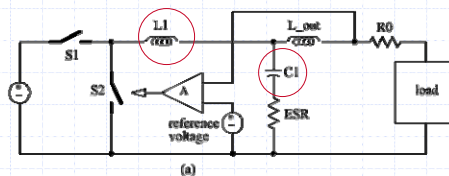
- The VRM converts one DC voltage to another, for example 5V to 1.8V.
- It has a **reference voltage** and a **feedback loop**.
- The bandwidth of the regulation loop is usually between 1kHz to several hundred kHz.
- At frequencies above the loop bandwidth, the VRM becomes **high impedance**.



Voltage Regulator Module (VRM) : working mechanism

Inductor

1. **S1 close, S2 open** : load is demanding current, L1 is storing energy.
2. If L1 supply more current than the load demanding, **S1 open, S2 close**.
3. Current continues to flow to the load until **S2 opens, S1 closes** again.

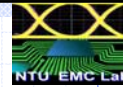


Capacitor

C1 smooths out the voltage.

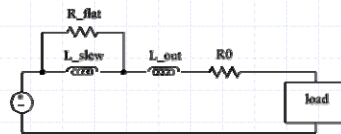
Amplifier A

1. When the load voltage is too low, it causes the switches and inductor to ramp up the current.
2. When the load voltage is too high, it causes the switches and inductor to ramp down the current.

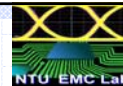


Voltage Regulator Module (VRM) : linear model

1. The buck regulator is **nonlinear** because switches open and closed as a function of time.
2. A **linear model** of VRM is necessary **in the design phase** of the Power Bus.
3. The linear model below is accurate enough to estimate the amount of the bulk capacitance.



1. **R0** is the value of the resistor between VRM sense point and the actual load.
2. **L_out** is the output inductance
 - It may be the **cables** that connect the VRM to the system board (200nH)
 - Or the inductance of the **pins** connecting the VRM to the CPU module (4nH)
3. **R_flat** is the ESR of the capacitor associated with the VRM.
4. **What is the L_slew and its value ?**

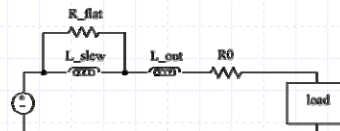


Voltage Regulator Module (VRM) : linear model

1. **L_slew** is the only element in the linear model that is not traceable back to an element in the nonlinear VRM model.
2. The value of **L_slew** is chosen so that current will be ramped up in the linear model in about the same time that it is ramped up in a real VRM.

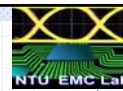
For example

$$L_{slew} = V \frac{dt}{di} = (1.8V * 0.05) \frac{15\mu Sec}{20A} = 67.5nH$$



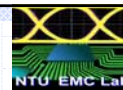
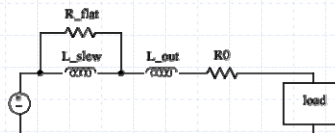
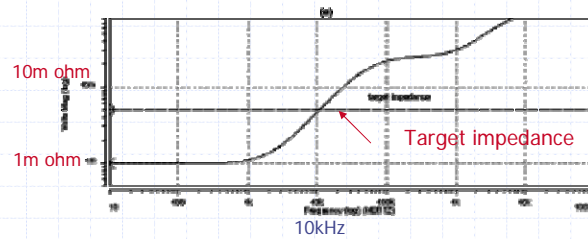
Typical parameters for a VRM attached on a processor module

parameter	value	units
R0	1	mOhm
L_out	4	nH
R_flat	30	mOhms
L_Slew	67.5	nH



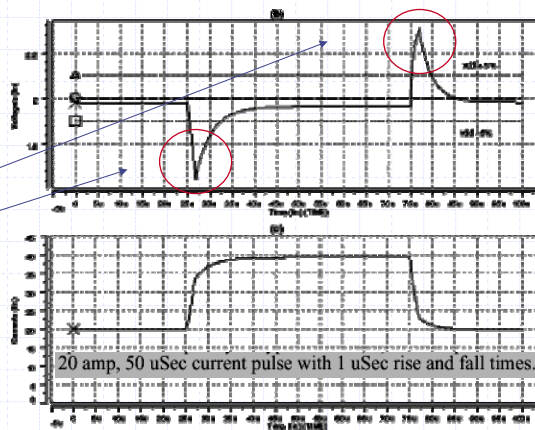
VRM : behavior example

Impedance V.S. Frequency

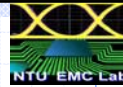


VRM : behavior example

VRM voltage v.s. time transient



Over spec. Why ?



Bulk Capacitance

Bulk capacitors are necessary to maintain the power bus impedance at frequencies above those maintained by the VRM and below the frequencies where ceramic capacitors are effective.

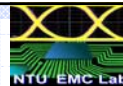
The capacitance value can be roughly estimated by

$$C = I \frac{\Delta t}{\Delta V}$$

Ex: For 20A current transient, the VRM responds in 15 μ s, and the PDS should remain in 5% of 1.8V.

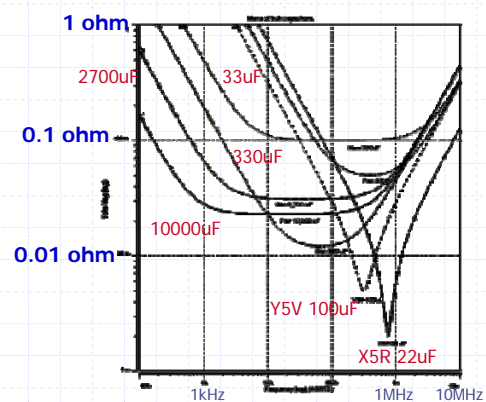
$$C = I \frac{dt}{dv} = 20 \text{ A} * \frac{15 \mu\text{s}}{1.8 \text{ V} * 0.05} = 3333 \mu\text{F}.$$

Does the capacitance value enough or over ?



Bulk Capacitance: frequency response

A: This estimation maybe overestimate the required bulk capacitance by a factor 2 because The VRM is ramping up current and the average current may be half of the final value during the VRM response time.



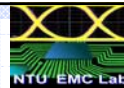
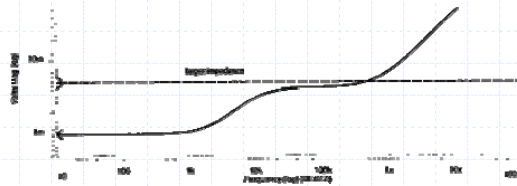
Bulk capacitance can be modeled by the series of RLC



Bulk Capacitance: behavior

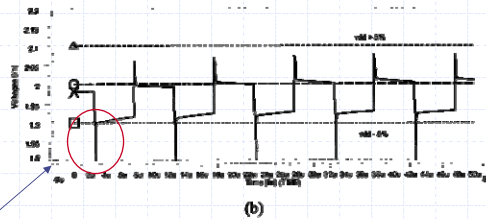
Output impedance v.s. frequency when a VRM is placed in parallel with 5 x 2700uF electrolytic bulk capacitance

The impedance is under spec up to almost 1MHz

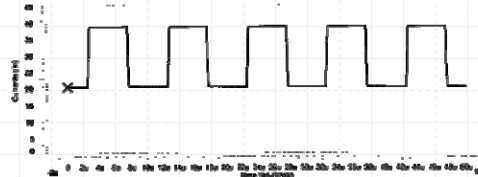


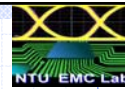
Bulk Capacitance: behavior

20A current transients at 100kHz with 200ns rise/fall time



Over spec (5%), why ?





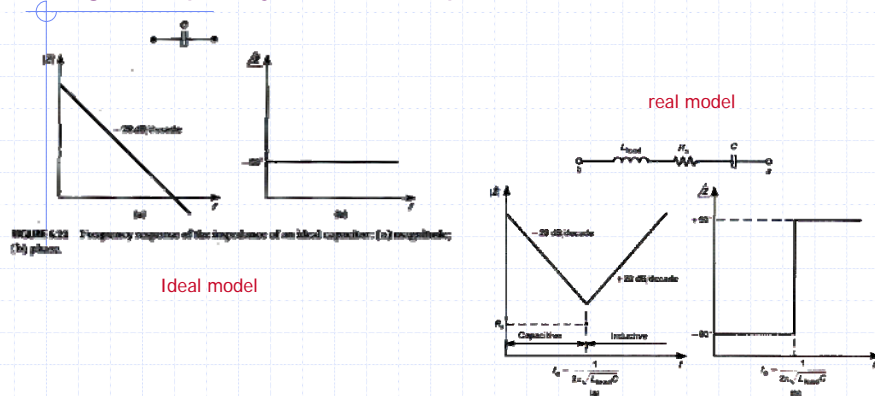
High Frequency Ceramic Capacitance: Type

Dielectric type	ESR	Reliability	Value
NPO	lowest	best	Up to a few nF
X7R	dependent	good	Several nF to several uF
X5R	dependent	fair	Several nF to 100 uF
Y5V	dependent	Poor	High capacitance

Package size:
0603
1206
0805



High Frequency Ceramic Capacitance: Model

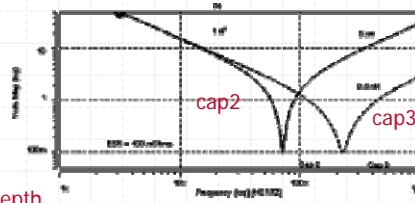
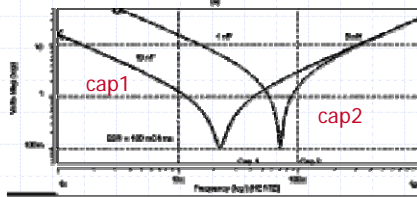




Ceramic Capacitance: Model behavior

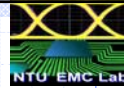
TABLE III
CAPACITOR VALUES ANALYZED IN FIG. 6

	cap 1	cap 2	cap 3
Capitance	10 nF	1 nF	1 nF
ESR	100 mOhm	100 mOhm	100 mOhm
Inductance	5 nH	5 nH	0.5 nH

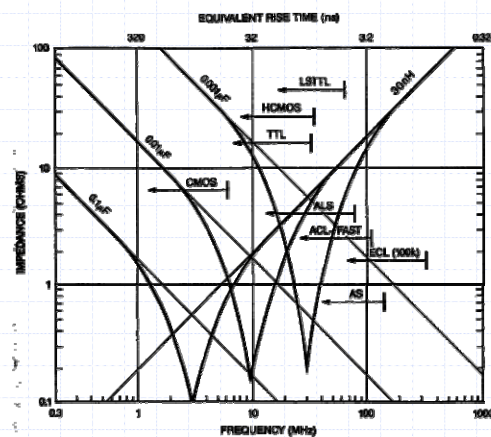


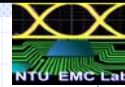
depth

The depth is decided by what ?
The resonance frequency is decided by what ?



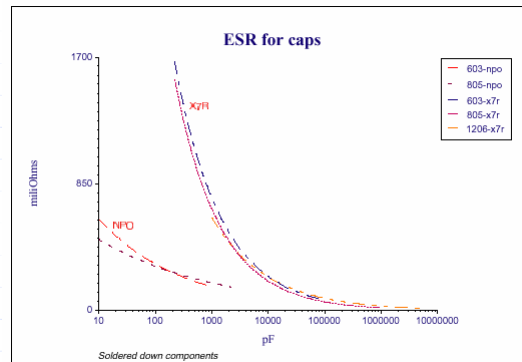
Ceramic Capacitance: Model behavior





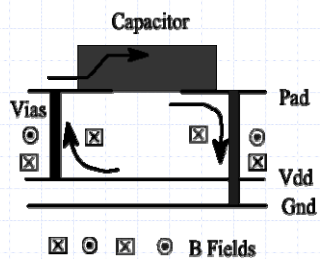
ESR of Ceramic Capacitance

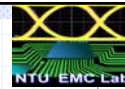
ESR can be measured by the HP4291A (Impedance Analyzer) by a SMA testing fixture
A better technique employing the low impedance head connected to the IA.



ESL of Ceramic Capacitance

Who contributes the ESL of the decoupling capacitor ??

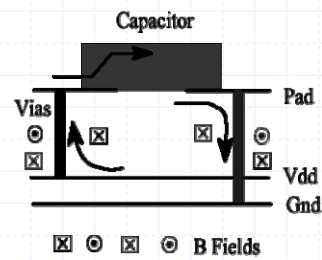




ESL of Ceramic Capacitance

There are THREE components of the ESL.

1. Pad layout
2. Capacitor height
3. Power plane spreading inductance

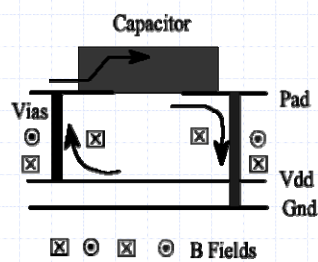


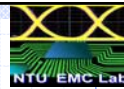
ESL of Ceramic Capacitance : Pad inductance

The pad layout consists of

1. Via placement with respect to the pad
2. The length and width of the traces connected to the pad
3. The way the via is connected to the power/ground planes

Which parts dominate the ESL of the pad layout ??





ESL of Ceramic Capacitance : Pad inductance

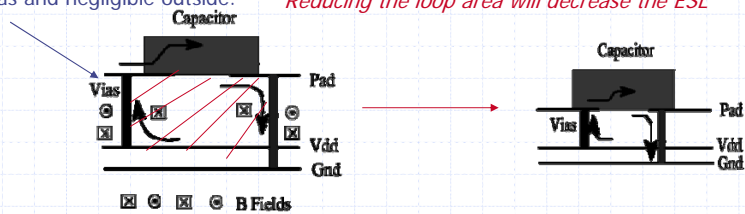
$$L \triangleq \frac{\varphi}{I}$$

$$\varphi = \oint \vec{B} \cdot d\vec{s}$$

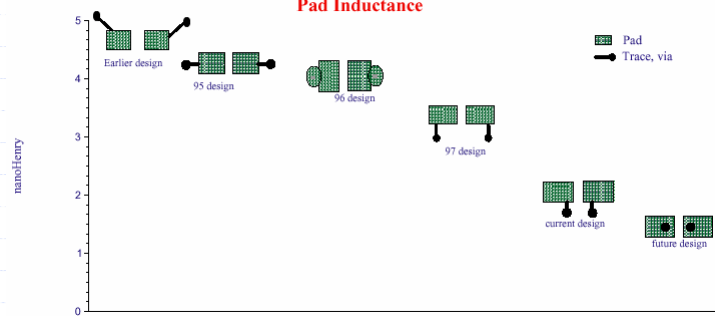
The location of the power/ground in the PCB stackup controls the height of the via which is the major contributor of the pad inductance.

Magnetic field concentrate between two vias and negligible outside.

Reducing the loop area will decrease the ESL



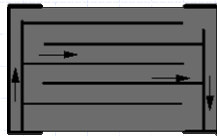
ESL of Ceramic Capacitance : Pad inductance





ESL of Ceramic Capacitance : Capacitance height

Capacitor height

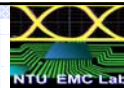


For thicker capacitor, the current has to flow up and down and effectively increase the length of the current loop.

Thickness(mils)	Inductance(pH)
20	300
30	450
40	600
50	700

← 0805

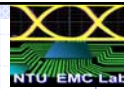
Figure 10: Height adds inductance to the capacitor.



ESL of Ceramic Capacitance : PWR/GND spreading inductance

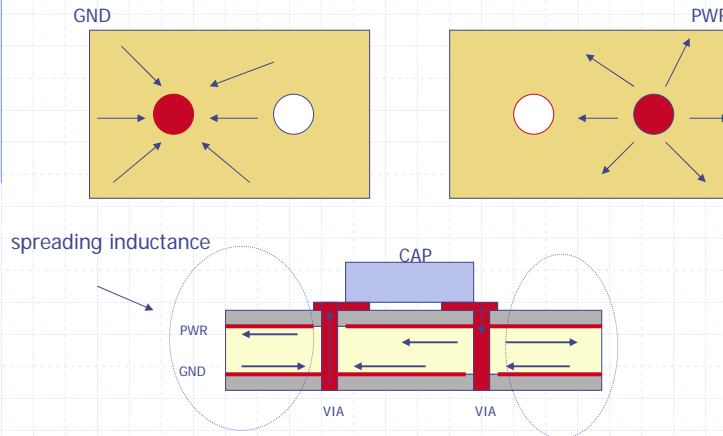
The last factor contributing to the ESL is the Spreading inductance between the PWR/GND.

What is the meaning of the spreading inductance ?

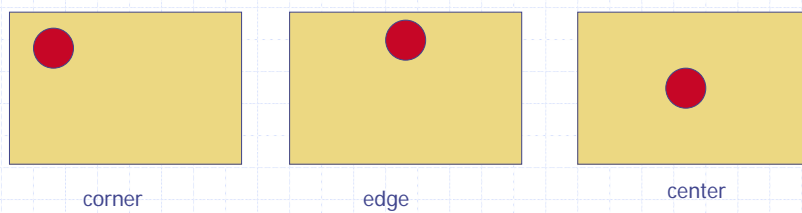


ESL of Ceramic Capacitance : PWR/GND spreading inductance

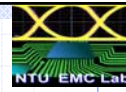
What is the meaning of the spreading inductance ?



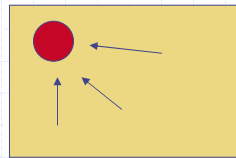
ESL of Ceramic Capacitance : PWR/GND spreading inductance



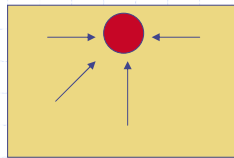
Which one has larger spreading inductance ? Why ?



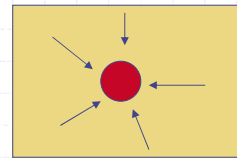
ESL of Ceramic Capacitance : PWR/GND spreading inductance



Corner (1.)



Edge (2.)

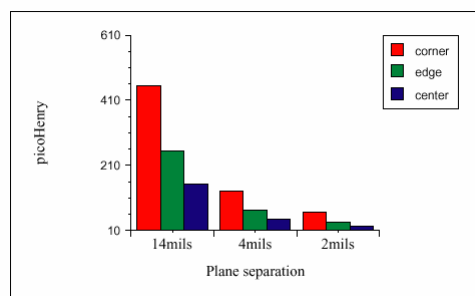


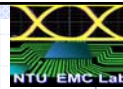
Center (3.)



ESL of Ceramic Capacitance : PWR/GND spreading inductance

Spreading inductance for different via locations for various plane separations

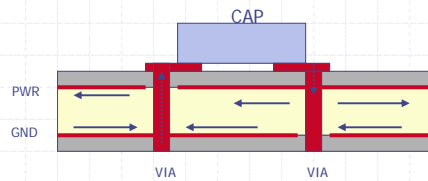




ESL of Ceramic Capacitance : Comparison

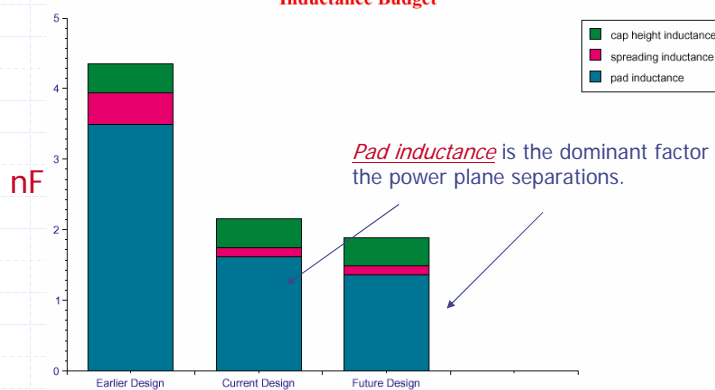
1. Pad layout
2. Capacitor height
3. Power plane spreading inductance

Which one is the main contributor of the ESL ??



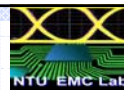
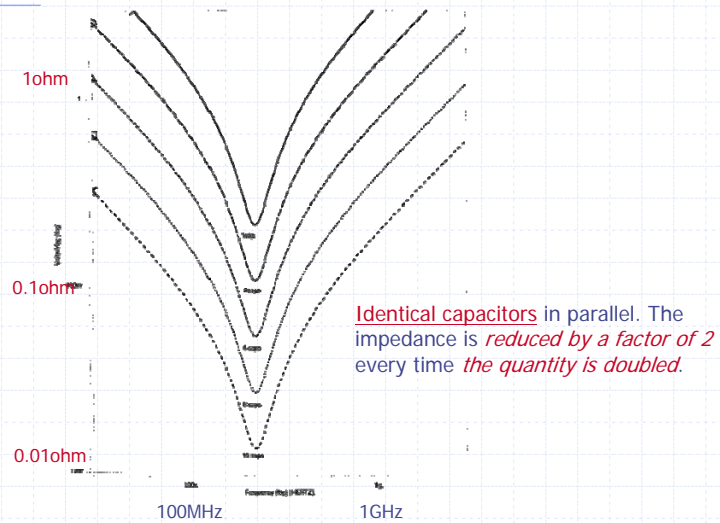
ESL of Ceramic Capacitance : Comparison

Inductance Budget

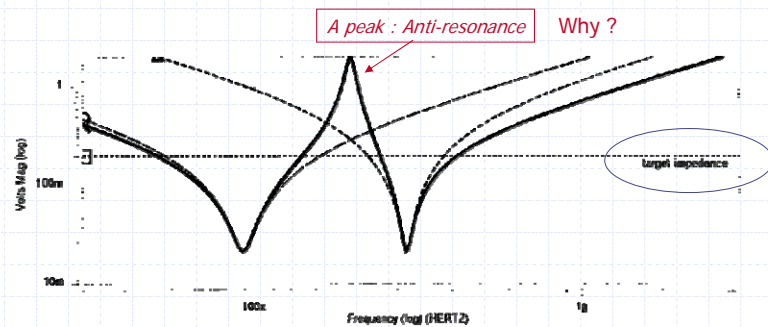


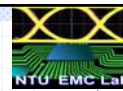


Shunting of Ceramic Capacitance : Identical capacitors

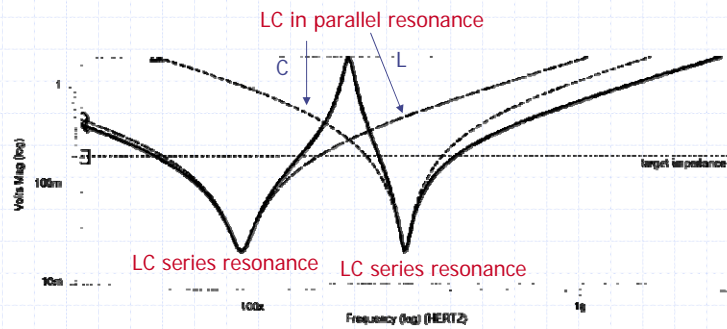


Shunting of Ceramic Capacitance: different values





Shunting of Ceramic Capacitance: different values

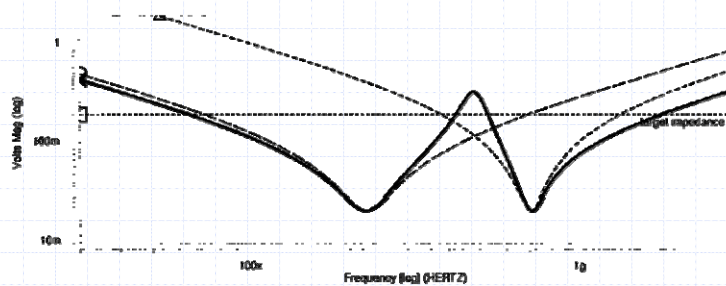


How to reduce the peak of the anti-resonance to meet the target impedance ??



Shunting of Ceramic Capacitance: different values

Minimizing the ESL is the most effective way to reduce the height of the peak.

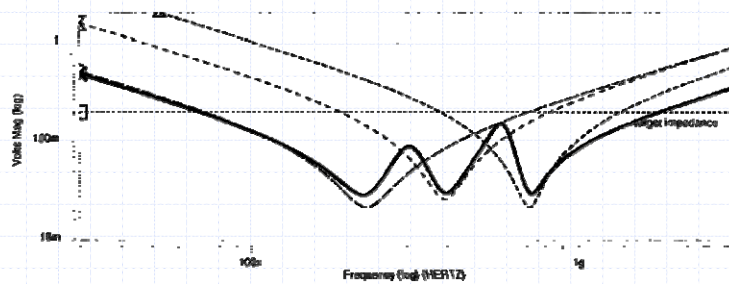




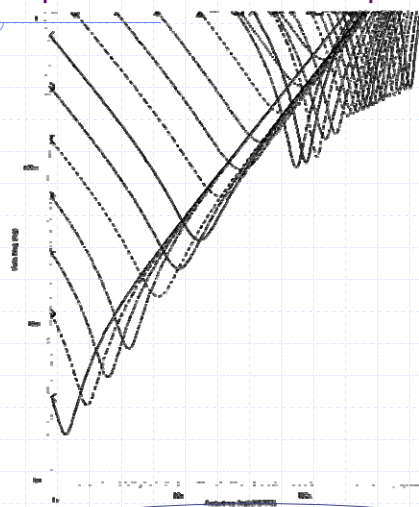
Shunting of Ceramic Capacitance: different values

The anti-resonance also becomes high if large gaps exist in capacitive value.

Three capacitors are used but with the total values being equal to the previous case.

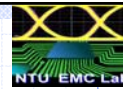


Impedance of Ceramic Capacitance

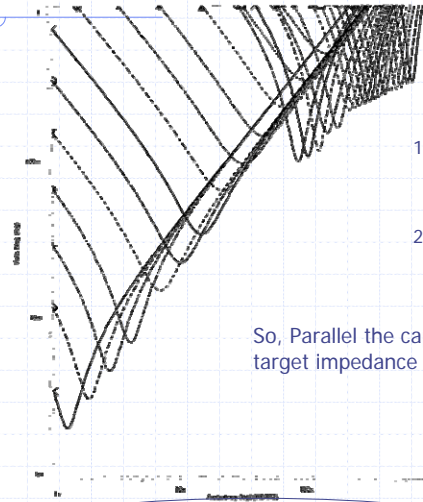


What capacitor behavior do you see ??

fig. 13: Impedance versus frequency for several X7R and NPO capacitors.



Impedance of Ceramic Capacitance



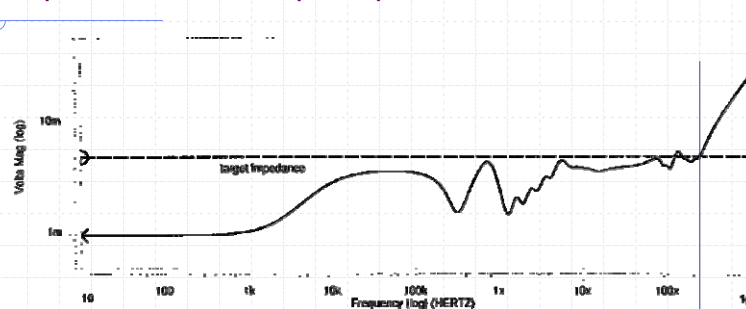
1. The resonance frequency goes up as the capacitance value decreases. Why? (ESL decreases)
2. The minimum impedance did not go down as the capacitance value decreases. Why? (ESR is lower for higher capacitive values)

So, Parallel the capacitor is necessary for meeting the target impedance at high frequencies.

fig. 13: Impedance versus frequency for several X7R and NPO capacitors.

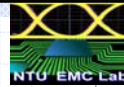


Impedance of a complete power bus



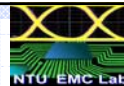
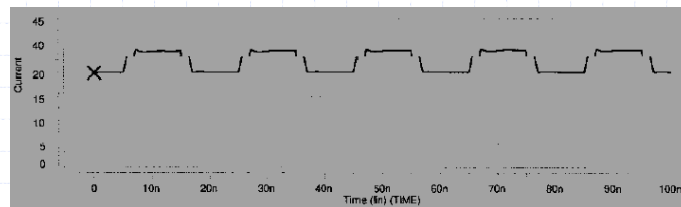
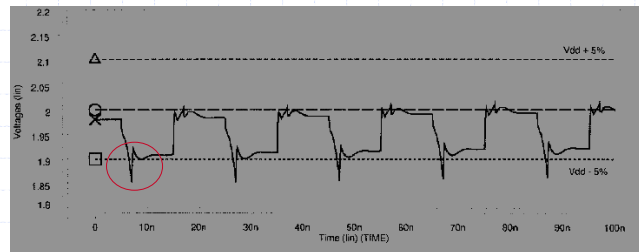
Impedance v.s. frequency of PDS with VRM, 7 bulk capacitors, 115 ceramic capacitors.

The PDS meets the target impedance up to 200MHz.



Impedance of a complete power bus

20Amp, 50MHz current transients that have 2ns rise time



Summary

*The target impedance is met in the frequency domain, noise
In time domain stays below a specified amount.*